

600V/30A IGBT IPM DIP-24H 三相全桥驱动智能功率模块

Features

- Integrated 6 low-loss IGBTs (600V/30A)
- Integrated high voltage gate drive circuit (HVIC)
- Built-in under voltage protection and over temperature, Over current protection and temperature output
- Insulation class 1500Vrms / min
- High reliability and thermal stability, good parameter consistency
- Built-in temperature output

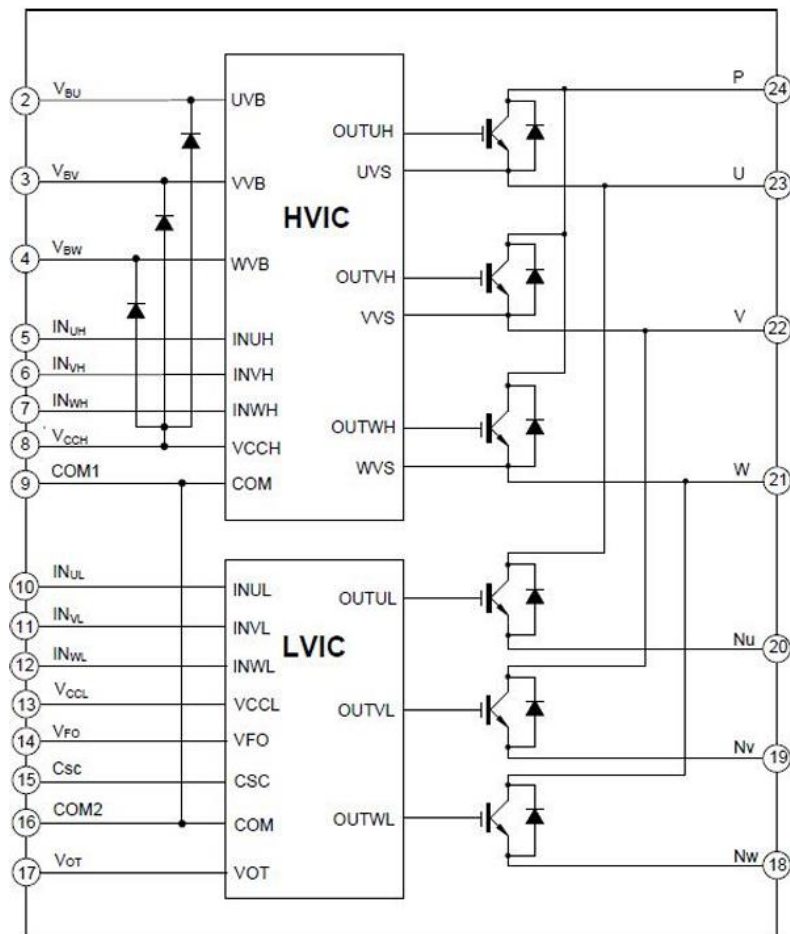


DIP-24H

Applications

- Airconditioning compressor
- Refrigerator compressor
- Low power inverters

Internal Electrical Schematic



Absolute Maximum Ratings $T_J = 25^\circ\text{C}$, unless otherwise noted

| Parameter | Symbol | Value | Unit |
|---|------------------|--------------|------|
| Inverter Section | | | |
| DC link supply voltage of P-N | V_{PN} | 450 | V |
| DC link supply voltage of P-N (surge) | $V_{PN(Surge)}$ | 500 | V |
| Collector-emitter voltage | V_{ce} | 600 | V |
| The collector continuous current of a single IGBT, $T_C=25^\circ\text{C}$ | I_C | 30 | A |
| The peak collector current of a single IGBT, $T_C=25^\circ\text{C}$, pulse width <1ms | I_{CP} | 60 | A |
| Maximum power dissipation per module collector, $T_C=25^\circ\text{C}$, $T_C=25^\circ\text{C}$ | P_C | 83 | W |
| Control section | | | |
| Control the supply voltage | V_{CC} | 20 | V |
| High-side control voltage | V_{BS} | 20 | V |
| Input signal voltage | V_{IN} | -0.3~VCC+0.3 | V |
| Fault output supply voltage | V_{FO} | -0.3~VCC+0.3 | V |
| Operating junction temperature range | T_J | -40 to 150 | °C |
| Working shell temperature range, $T_J \leq 150^\circ\text{C}$ | T_C | -30 to 100 | |
| Storage temperature range | T_{STG} | -40 to 125 | °C |
| IGBT crusts thermal resistance | $R_{\theta JCB}$ | 1.5 | °C/W |
| FRD crusts thermal resistance | $R_{\theta JCF}$ | 2.35 | °C/W |
| Isolation test voltage (1min, RMS, f = 60Hz) | V_{ISO} | 1500 | Vrms |

Note 1: The maximum junction temperature of the power chip is 150°C , in order to ensure that IPM can work safely, it is recommended that the average junction temperature $T_J \leq 125^\circ\text{C}$ ($\theta_{TC} \leq 100^\circ\text{C}$)

Recommended Operation Conditions $T_J = 25^\circ\text{C}$, unless otherwise noted

| Control section | | | | | |
|-------------------------------|----------|------|------|------|------|
| Parameter | Symbol | Min. | Typ. | Max. | Unit |
| Busbar voltage between PNs | V_{PN} | - | 300 | 400 | V |
| Control the supply voltage | V_{CC} | 13.2 | - | 20 | V |
| High-side control voltage | V_{BS} | 13.0 | - | 20 | V |
| Input signal voltage | V_{IN} | VSS | - | VCC | V |
| High-side gate output voltage | V_{HO} | VS | - | VB | V |
| Low-side gate output voltage | V_{LO} | VSS | - | VCC | V |

Electrical Characteristics $T_J = 25^\circ\text{C}$, unless otherwise noted

| Inverter Section | | | | | | |
|--------------------------------------|---------------|--|------|------|------|---------------|
| Parameter | Symbol | Condition | Min. | Typ. | Max. | Unit |
| Collector-emitter saturation voltage | $V_{CE(SAT)}$ | $V_{CC}=V_{BS}=15\text{V}$, $V_{IN}=5\text{V}$ $I_C=30\text{A}$, $T_J = 25^\circ\text{C}$ | - | 1.7 | 2.2 | V |
| FRD forward voltage | VF | $V_{IN}=0\text{V}$, $I_F=30\text{A}$, $T_J = 25^\circ\text{C}$ | | 1.5 | 2.0 | V |
| Switching time (high side) | t_{on} | $V_{PN} = 300\text{V}$, $V_{CC} = V_{BS} = 15\text{V}$, $I_C = 30\text{A}$, $V_{IN} = 0\text{V} \leftrightarrow 5\text{V}$, The inductive load is detailed in Figure 1 | - | 783 | - | ns |
| | t_r | | - | 62 | - | ns |
| | t_{off} | | - | 722 | - | ns |
| | t_f | | - | 48 | - | ns |
| | t_{rr} | | - | 110 | - | ns |
| Switching time (low side) | t_{on} | | - | 946 | - | ns |
| | t_r | | - | 175 | - | ns |
| | t_{off} | | - | 790 | - | ns |
| | t_f | | - | 48 | - | ns |
| | t_{rr} | | - | 123 | - | ns |
| Collector-emitter current | I_{CES} | $V_{CE}=600\text{V}$ | - | - | 250 | μA |

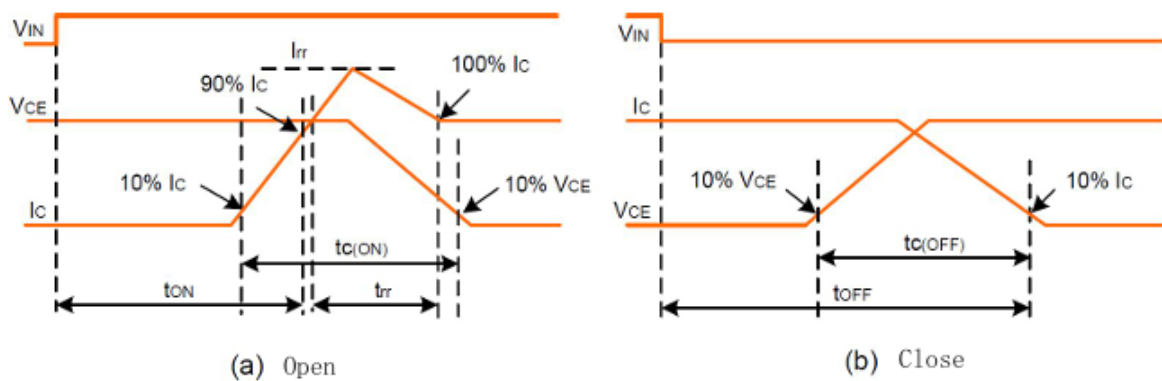
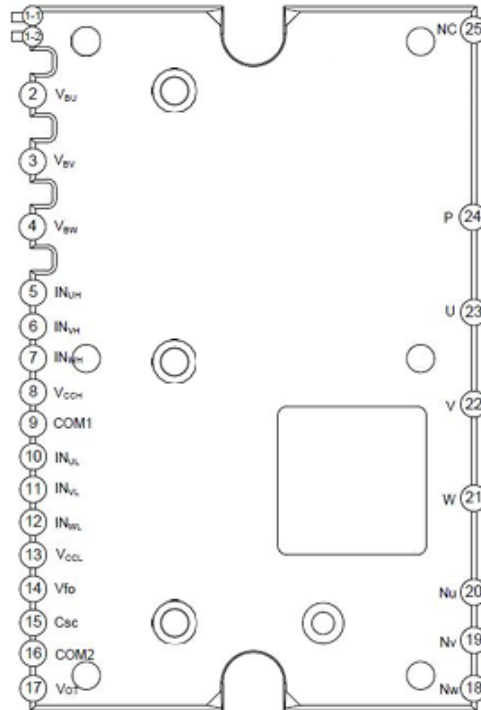


Figure 1 Definition of switching time

| Control section | | | | | | |
|--|---------------|--|-------|------|-------|-------------|
| Parameter | Symbol | Condition | Min. | Typ. | Max. | Unit |
| Quiescent VCC supply current | I_{QCC} | VCC=15V, VIN=0V | - | - | 3.5 | mA |
| Quiescent VBS supply current | I_{QBS} | VBS=15V, VINH=0V | - | 75 | - | uA |
| Fault output voltage | VFOH | VSC=0V, VFO pulls up 10K Ω Resistor to 5V | 4.9 | - | - | V |
| | VFOL | VSC=1V, IFo=1mA | - | - | 0.9 | V |
| Fault output pulse width | t_{FO} | Fault duration | 40 | - | - | us |
| Short-circuit protection trigger voltage | $V_{SC(ref)}$ | VCC=15V | 0.415 | 0.46 | 0.505 | V |
| Over-temperature protection | TSD | LVIC temperature | 100 | 120 | 140 | $^{\circ}C$ |
| Over-temperature protection hysteresis | ΔTSD | LVIC Hysteresis temperature | - | 10 | - | $^{\circ}C$ |
| Temperature output (Figure 2) | VOT | LVIC=25 $^{\circ}C$ | 0.88 | 1.13 | 1.39 | V |
| | | LVIC=90 $^{\circ}C$ | 2.63 | 2.77 | 2.91 | V |
| Low-side undervoltage protection (Figure 5) | UVCCCT | VCC senses the voltage | 10 | 11 | 12 | V |
| | UVCCR | VCC reset voltage | 9 | 10 | 11 | V |
| High-side undervoltage protection (Figure 6) | UVBST | VBS senses voltage | 10 | 11 | 12 | V |
| | UVBSR | VBS reset voltage | 9 | 10 | 11 | V |
| On-threshold voltage | VIH | Logic high | - | - | 2.5 | V |
| Shutdown threshold voltage | VIL | Logic low | 0.8 | - | - | V |

| Bootstrap diode section | | | | | | |
|--------------------------------|----------|----------------------------|------|------|------|------|
| Parameter | Symbol | Condition | Min. | Typ. | Max. | Unit |
| Forward voltage | V_F | IF=50mA, TC=25 $^{\circ}C$ | - | 2.5 | 4.0 | V |
| Reverse recovery time | t_{rr} | IF=10mA, TC=25 $^{\circ}C$ | - | 50 | - | ns |

Pin Assignment



Pin Description

| Pin Number | Pin name | Pin Description |
|------------|----------|--|
| 1-1 | COM | Internal common ground terminals, no connection |
| 1-2 | VCC | Internal power terminals, no connection |
| 2 | VBU | The U-phase high-side IGBT drives the floating supply voltage |
| 3 | VBV | The high-side IGBT of the V phase drives the floating supply voltage |
| 4 | VBW | The high-side IGBT of the W phase drives the floating supply voltage |
| 5 | INUH | U-phase high-side signal input |
| 6 | INVH | V-phase high-side signal input |
| 7 | INWH | High-side signal input of the W phase |
| 8 | VCCH | High side gate drive supply voltage |
| 9 | Com1 | Module Commons |

| Pin Number | Pin name | Pin Description |
|------------|----------|--|
| 10 | INUL | U-phase low-side signal input |
| 11 | INVL | V-phase low-side signal input |
| 12 | INWL | W-phase low-side signal input |
| 13 | VCCL | low-side gate drives the supply voltage |
| 14 | VFO | Fault output |
| 15 | Csc | External capacitors for short-circuit current sense input and low-pass filtering |
| 16 | Com2 | Module Commons |
| 17 | VOT | Temperature output |
| 18 | NW | W-phase DC negative terminal |
| 19 | NV | V-phase DC negative terminal |
| 20 | NU | U-phase DC negative terminal |
| 21 | W | W output |
| 22 | V | V output |
| 23 | U | U output |
| 24 | P | DC positive terminal |
| 25 | NC | No connection |

Description of the temperature output function

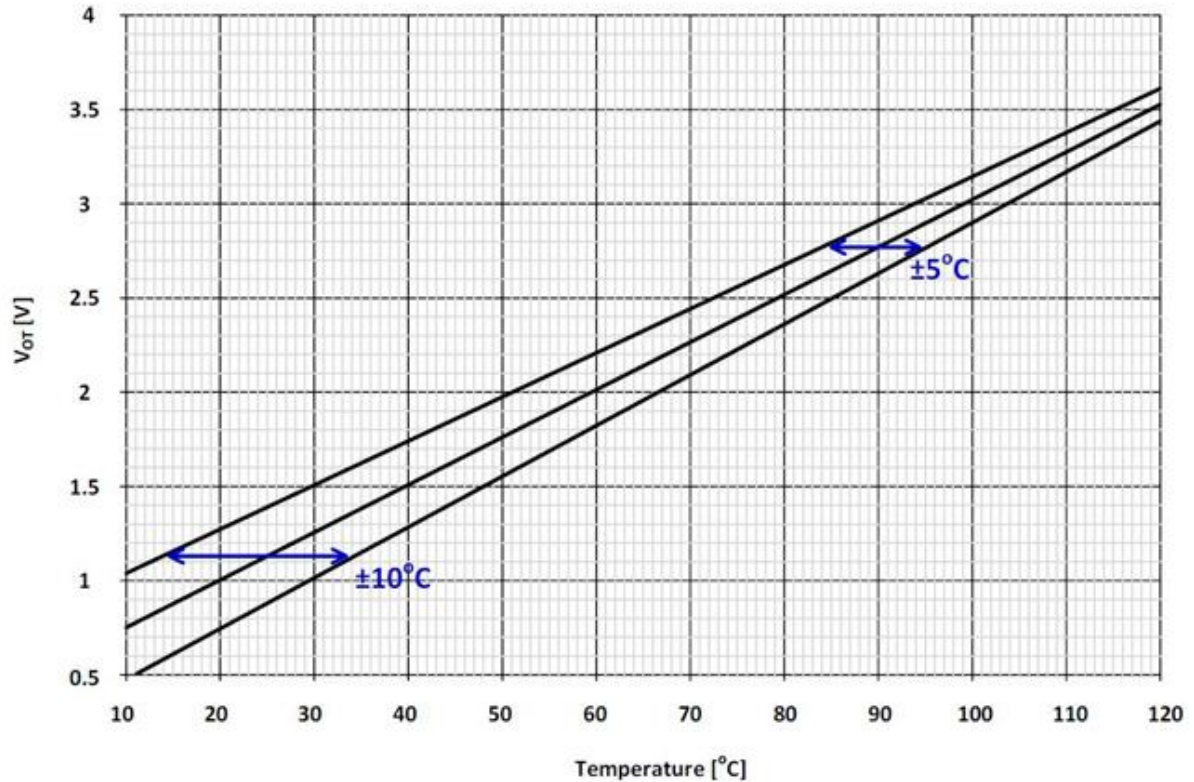


Figure 2 LVIC temperature VOT temperature characteristics

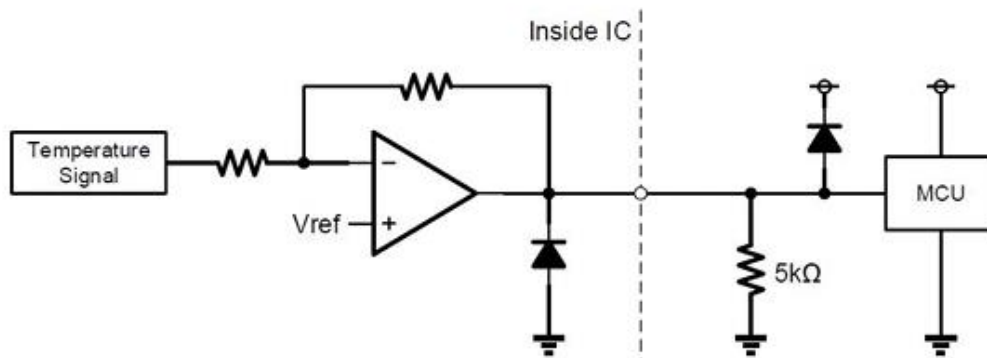


Figure 3 VOT output circuit

(1) If the temperature monitoring function is used, connect 5kΩ to the VOT pin, and ignore the internal OTP function. If the internal overtemperature shutdown function is used, keep the VOT pin on (no connection).

(2) When IPM is applied in low-voltage control (such as MCU working voltage of 3.3V), the output voltage of VOT may be greater than the control supply voltage of 3.3V in the case of a sharp rise in temperature, if the system is used for low-voltage control, it is recommended to connect a clamping diode between the control power supply and the VOT output signal to prevent overvoltage damage.

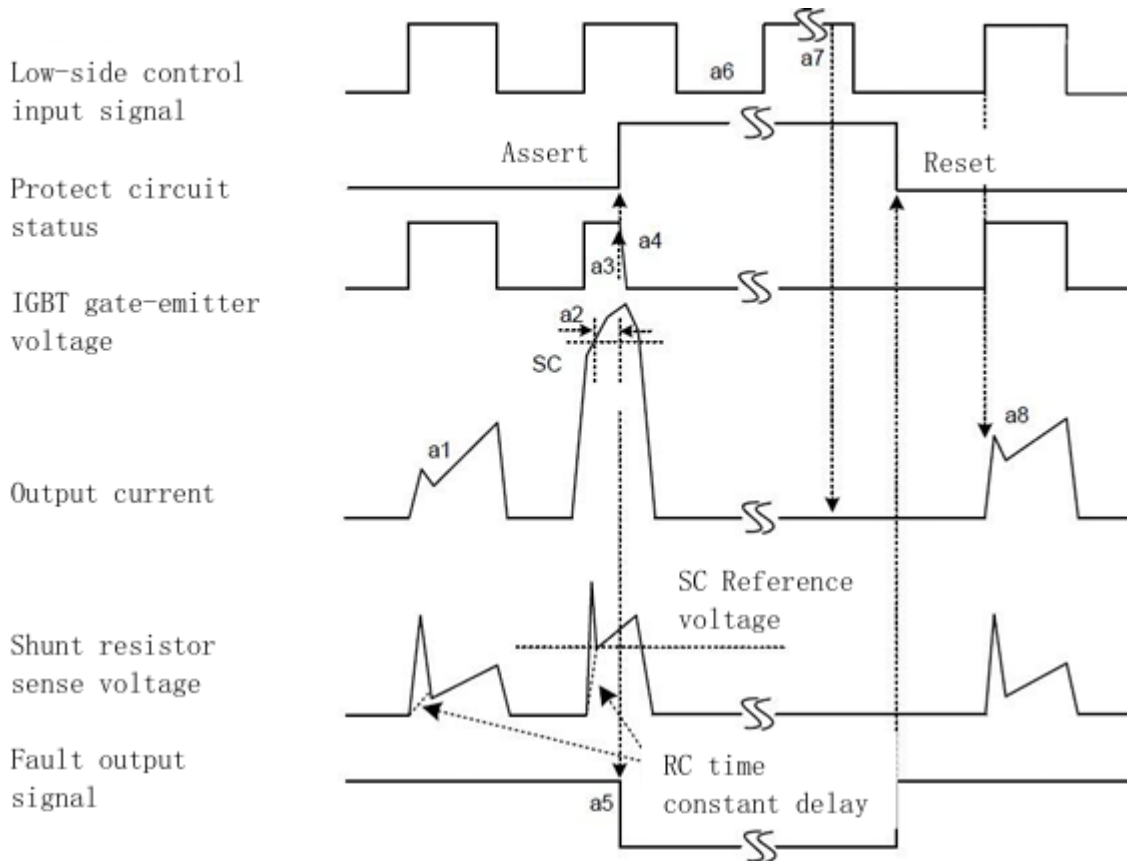


Figure 4.Short-circuit current protection (low side only)

(Includes external shunt resistor and RC connection).

a1: Normal operation: IGBT conducts and supplies current to the load.

a2: Short-circuit current detection (short-circuit triggering).

a3: All low-side IGBT gate hard interrupts.

a4: All low-side IGBTs are turned off.

a5: The fault output pin outputs a fixed pulse width signal ($t_{FO} \geq 40\mu s$).

a6: Input is "L": IGBT shutdown state.

a7: Input is "H": Although the input is "H", the IGBT is still in the off state during this time if there is a fault output signal.

a8: Normal operation: IGBT is on, current is supplied to the load.

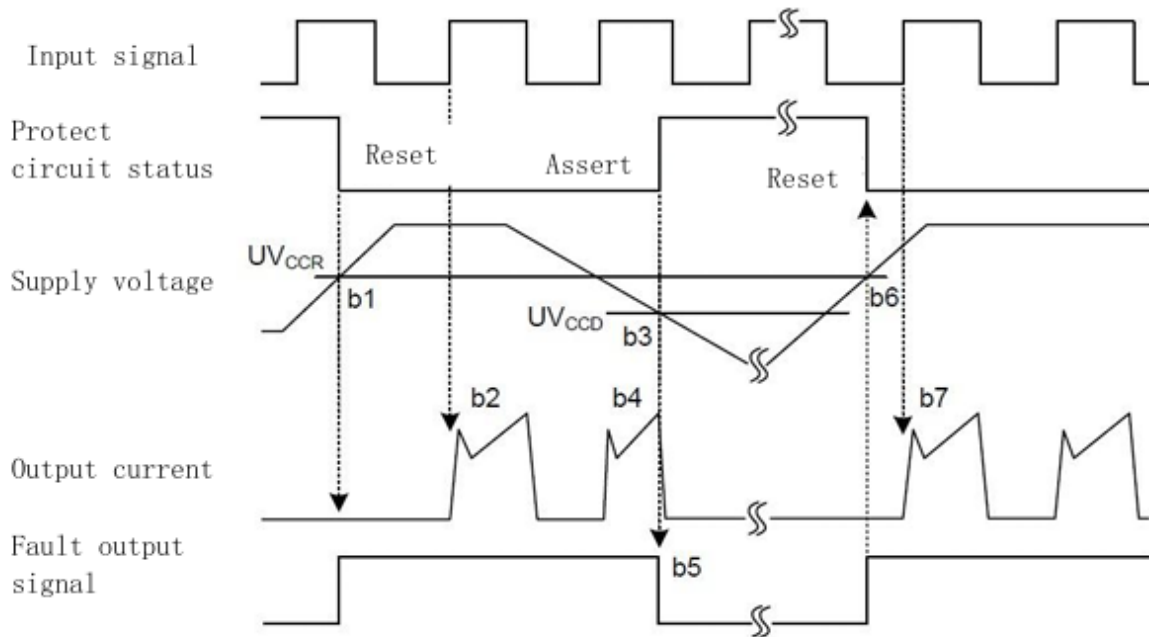


Figure 5: Undervoltage Protection (Low Side)

b1: The supply voltage rises to UV_{CCR} and the circuit starts working when the next input waveform arrives.

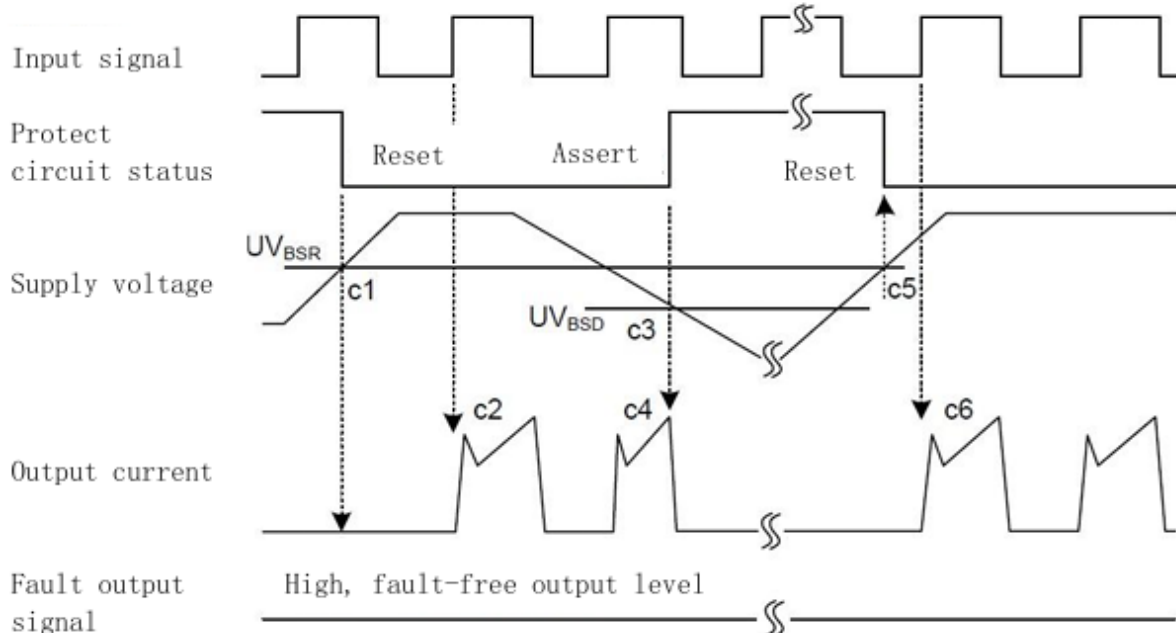
b2: Normal operation: IGBT conducts and supplies current to the load.

b3: Undervoltage detection point (UV_{CCD}).

b4: All low-side IGBTs are turned off regardless of the signal input.

b5: The FO pin outputs a fault signal ($t_{FO} \geq 40\mu s$) and continuously outputs a fault signal during undervoltage. b6: Undervoltage reset point (UV_{CCR}).

b7: normal operation: IGBT conducts and supplies current to the load.



- c1: The supply voltage rises to UVBSR, and the circuit starts working when the next input signal arrives.
- c2: Normal operation: IGBT conducts and supplies current to the load.
- c3: Undervoltage detection point (UVBSD).
- c4: IGBT is turned off regardless of signal input, but there is no fault signal output.
- c5: Undervoltage reset point (UVBSR).
- c6: normal operation: IGBT conducts and supplies current to the load.

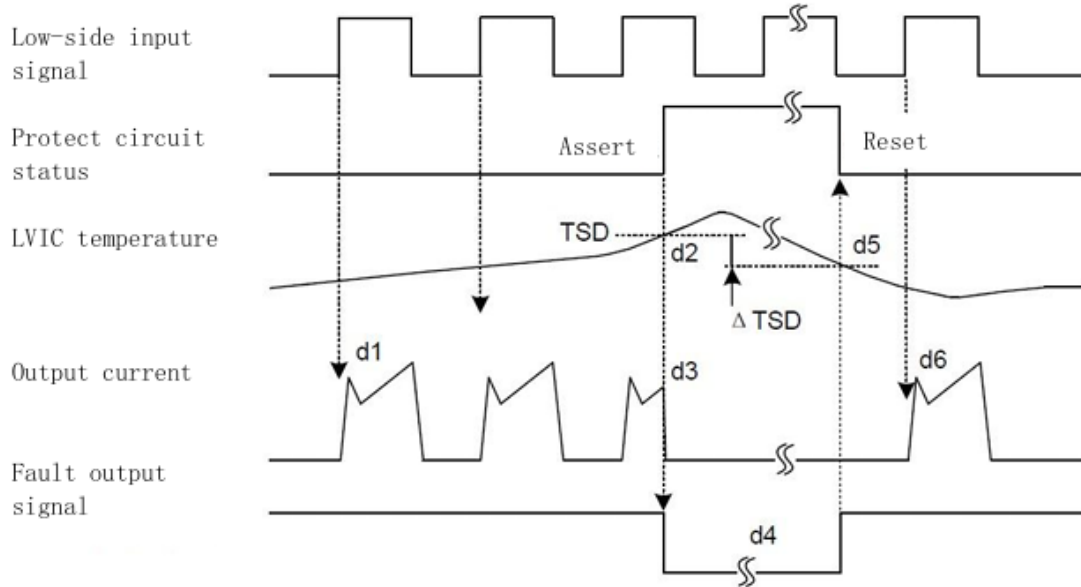


Figure 7. Overtemperature protection (low side only)

- d1: Normal operation: IGBT conducts and supplies current to the load.
- d2: LVIC temperature exceeds overtemperature protection trigger point (TSD).
- d3: All low-side IGBTs are turned off, regardless of the signal input.
- d4: Continuously outputs fault signals during overtemperature, and the minimum pulse width is 40us.
- d5: LVIC temperature will reset when the temperature falls below the overtemperature protection point.
- d6: When the next input signal control signal comes, the circuit enters normal working state.

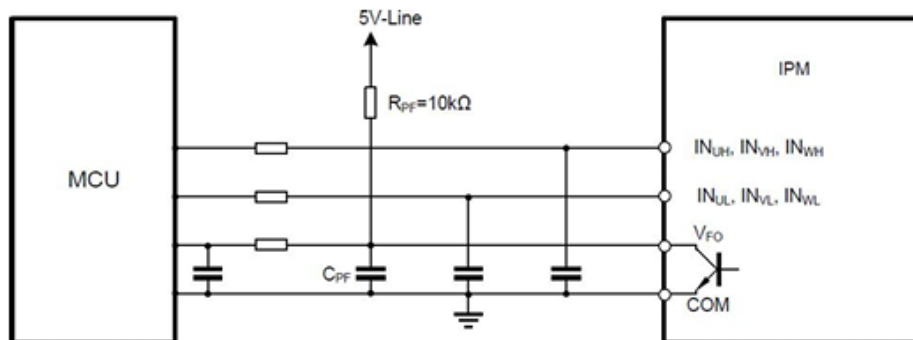
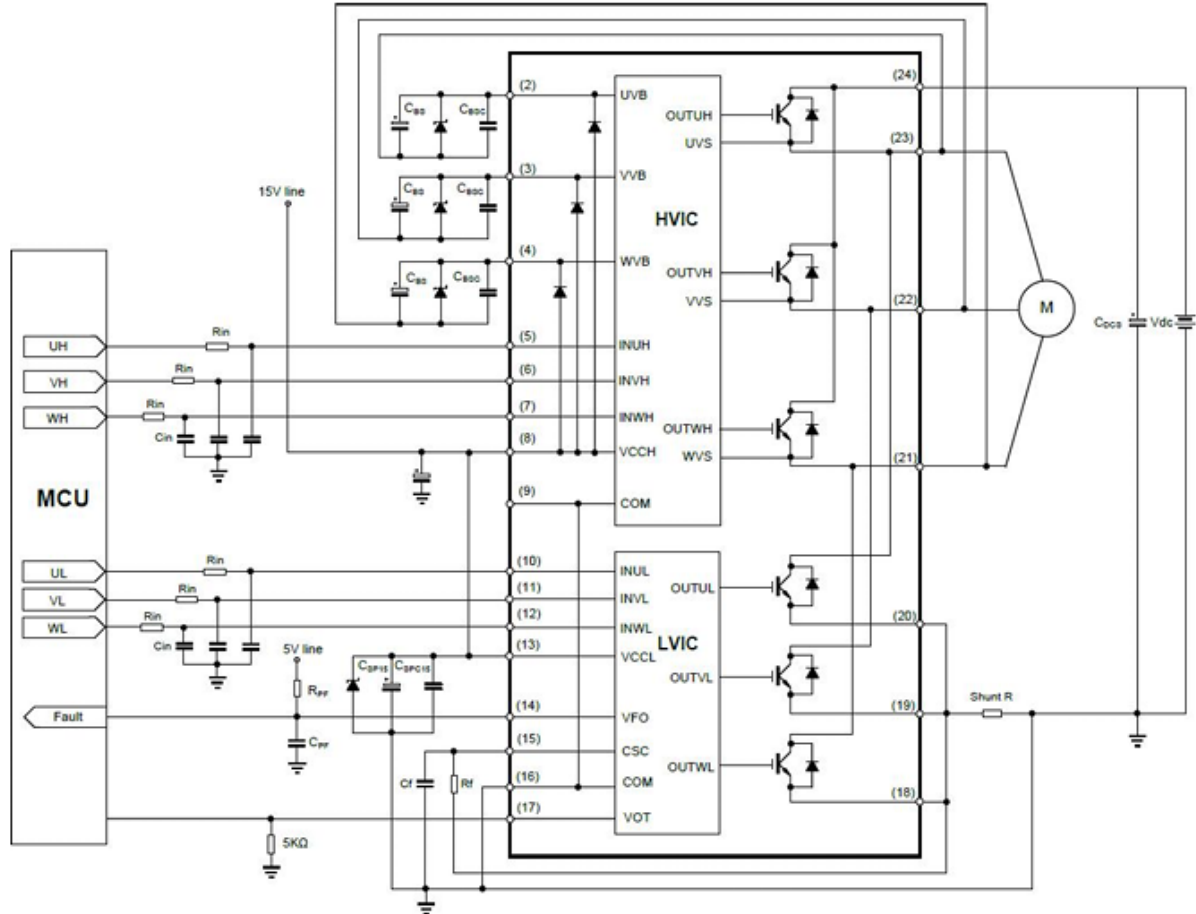


Figure 8. MCU input/output connection circuit (recommended)

Note: The RC coupling at each input should be adapted to the PWM control scheme and PCB layout. A 5K pull-down resistor is built into the IPM input signal section, so pay attention to the voltage drop at the input when using an external filter circuit.

Typical application circuit diagram



Remark:

- (1) The connection of each input pin should be as short as possible, otherwise it may cause misoperation;
- (2) The input signal is active high, and a 5 K Ω pull-down resistor is connected to ground at the input of each HVIC channel; In addition, an RC filter circuit can be added at the input to prevent surge noise caused by incorrect input;
- (3) In order to prevent surge damage, it is recommended to add a high-frequency non-inductive flat capacitor (0.1 μ F ~ 0.22 μ F) between PNs, and the connection of the capacitor should be made Keep it as short as possible;
- (4) The connection between the current sense resistor and the IPM should be as short as possible, otherwise the large surge voltage generated by the connection inductor may cause damage;
- (5) The filter capacitor at the input of the 15V power supply is recommended to be at least 7 times that of the bootstrap capacitor CBS;
- (6) Each external capacitor should be placed as close as possible to the IPM pin;
- (7) The VFO output is open and should be pulled up to the 5V supply through the resistor so that the Ifo is 1mA;
- (8) In the short-circuit protection circuit, please select Rf and CSC with time constants in the range of 1.5~2 μ s, and the wiring around Rf and CSC should be as short as possible. The Rf wiring should be close to the shunt resistor.

Package outline drawing

